

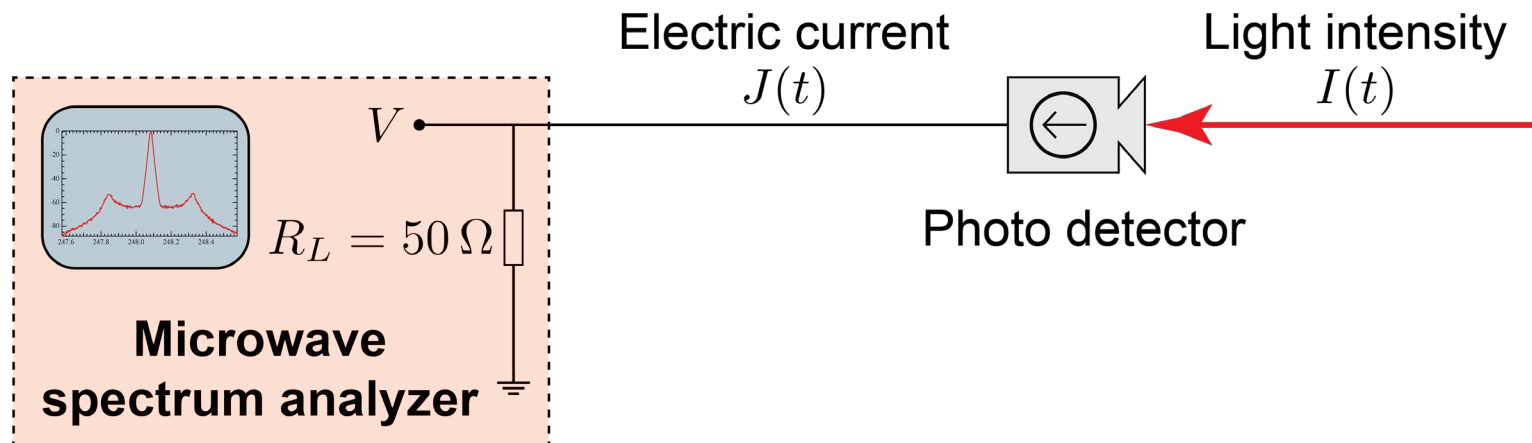
Ultrafast Laser Physics

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Chapter 10: Noise





The microwave spectrum analyzer measures the **power spectral density**

$$S_I(\omega) \propto \left| \tilde{I}(\omega) \right|^2 = F \{ \text{corr} (I(t), I(t)) \} = F \{ R(\tau) \}$$

autocorrelation function: $R(-\tau) = R(\tau)$

$$R(-\tau) \equiv \int I(t)I(t - \tau)dt = \text{corr} (I(t), I(t)) \xrightarrow{F} \tilde{I}(\omega)\tilde{I}^*(\omega)$$

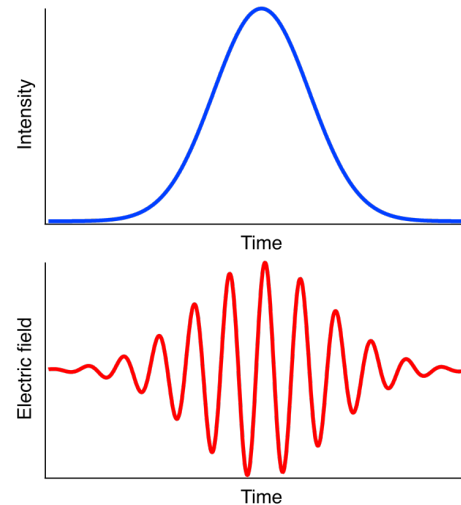
$$\tilde{I}(\omega) = F\{I(t)\} \propto F\{|E(t)|^2\}$$

slowly varying

$$\neq |E(\omega)|^2 = |F\{E(t)\}|^2$$

fast oscillating

$$S_I(\omega) = |\tilde{I}(\omega)|^2 \propto |F\{|E(t)|^2\}|^2$$

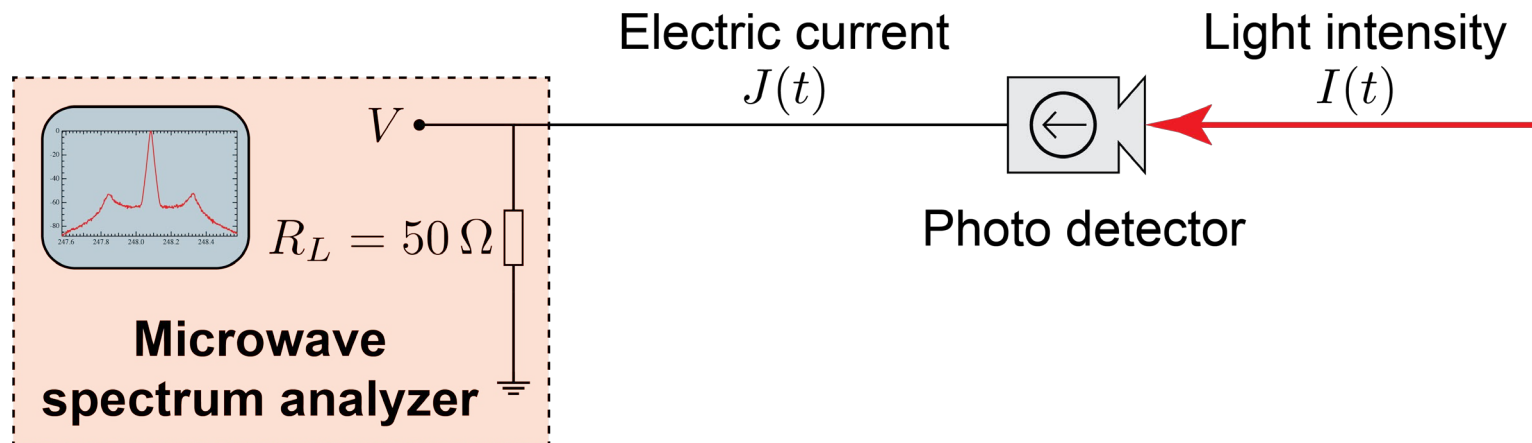


$$|\tilde{I}(\omega)|^2 = \tilde{I}(\omega)\tilde{I}^*(\omega) = F\{\text{corr}(I(t), I(t))\}$$

Wiener-Khinchin theorem

This is the autocorrelation function for the entire pulse train, not just a single pulse (in contrast to the intensity autocorrelation technique). I.e., time variable spans from $-\infty$ to $+\infty$.



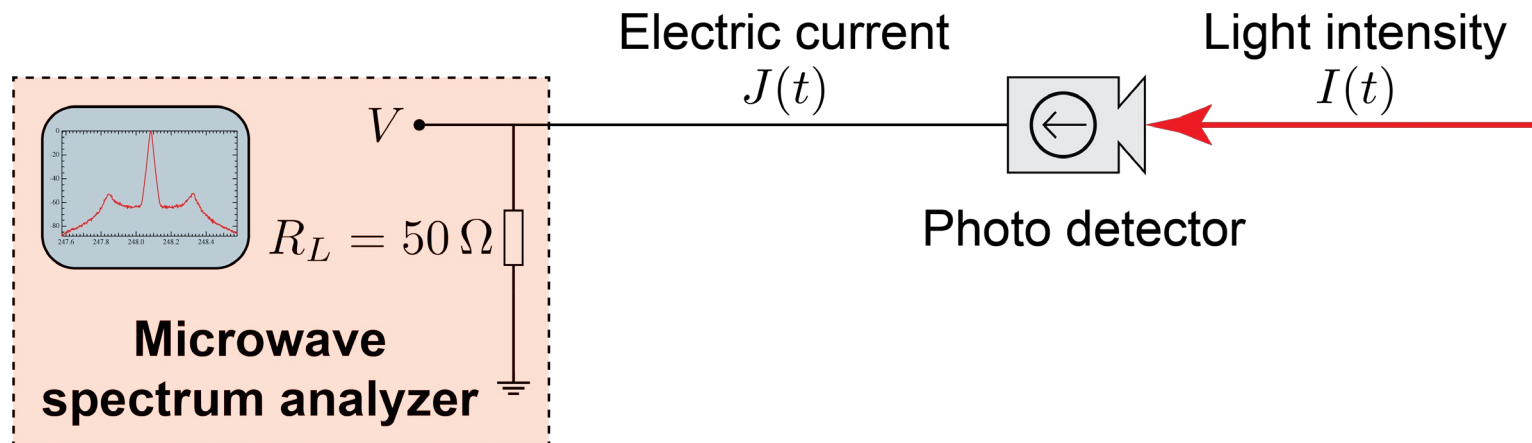


Modelocked laser without any noise

Pulses like delta-functions

$$I(t) = I_0 T \sum_{n=-\infty}^{+\infty} \delta(t - nT) \quad \longrightarrow \quad \tilde{I}(\omega) = F\{I(t)\} = 2\pi I_0 \sum_{n=-\infty}^{+\infty} \delta(\omega - n\omega_T)$$

For more details see: U. Keller et al., *IEEE J. Quantum.Electron.*, **25**, 280 (1989)

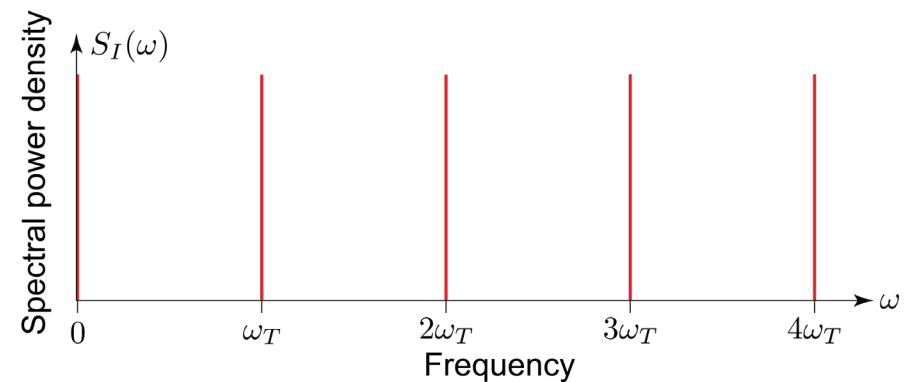


We measure the
power spectral density

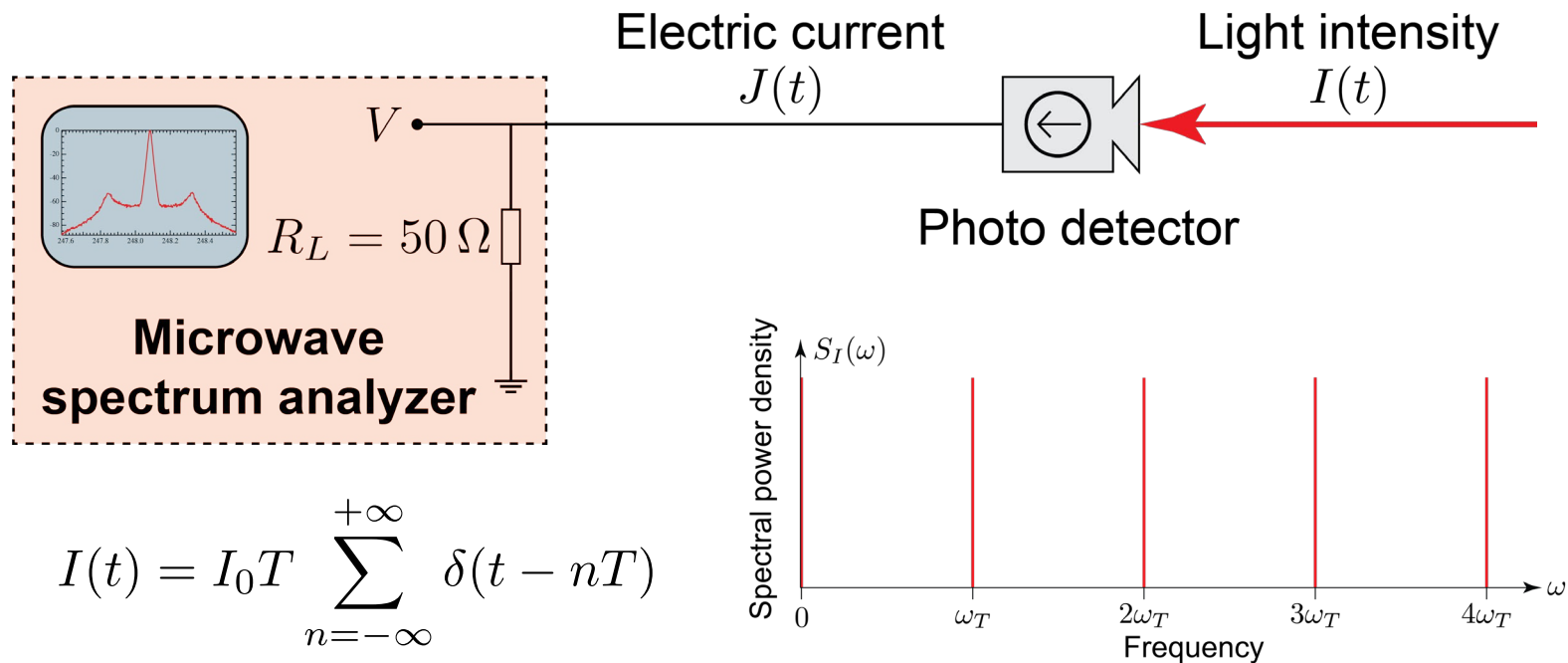
$$S_I(\omega) \propto |\tilde{I}(\omega)|^2$$

$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{m=-\infty}^{+\infty} \delta(\omega - m\omega_T)$$

$$\omega_T = \frac{2\pi}{T}$$



modelocked laser without any noise and any bandwidth limitations of microwave spectrum analyzer



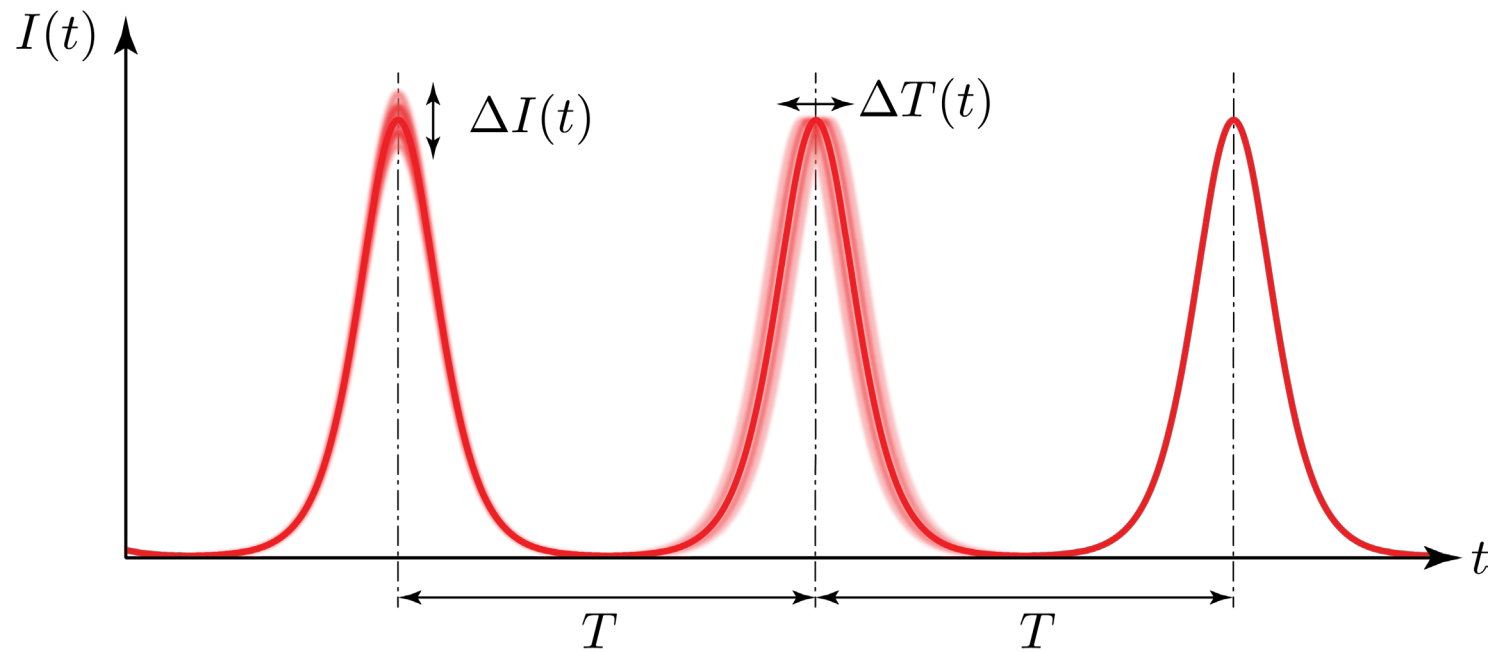
$$I(t) = I_0 T \sum_{n=-\infty}^{+\infty} \delta(t - nT)$$

$$I(t) = I_0 T (1 + N(t)) \sum_{n=-\infty}^{+\infty} \delta(t - nT - \Delta T(t))$$

normalized intensity noise

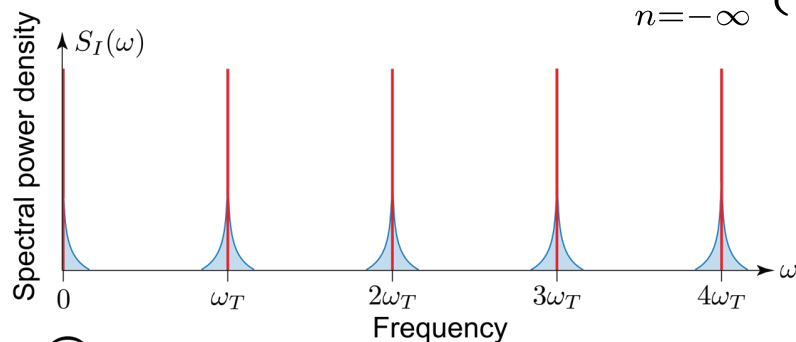
timing jitter

For more details see: U. Keller et al., *IEEE J. Quantum.Electron.*, **25**, 280 (1989)

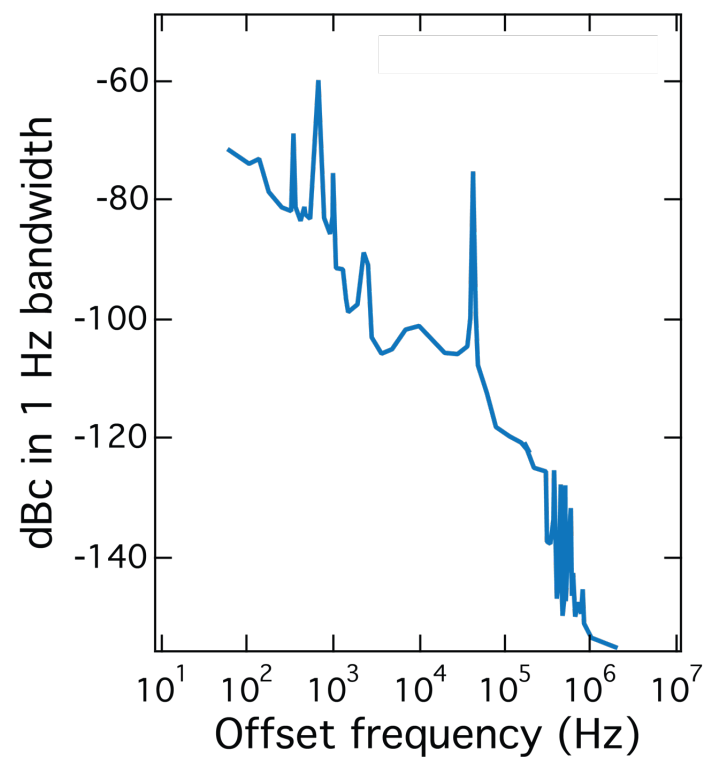
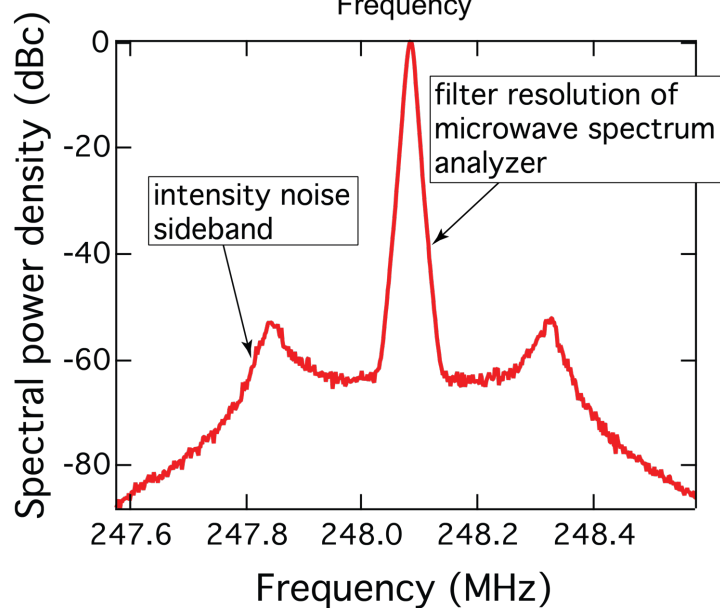


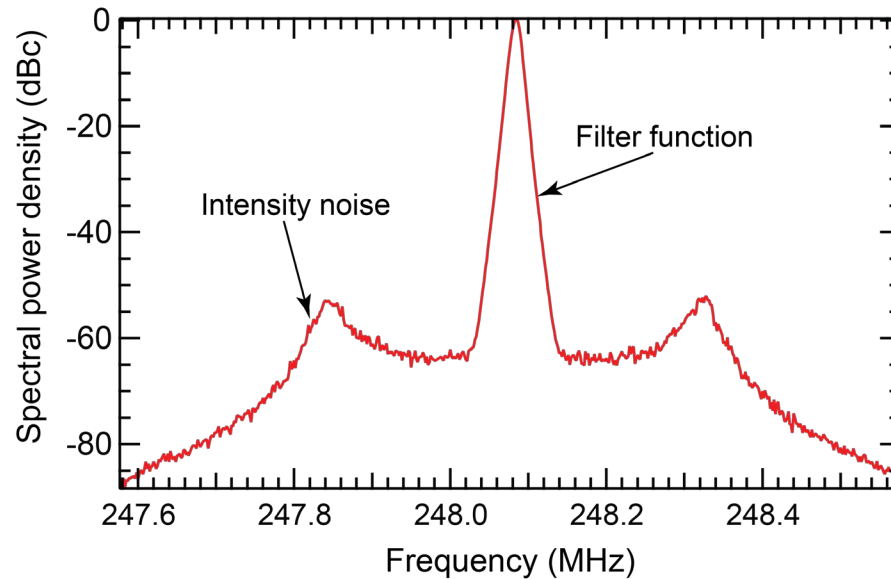
$$\Delta I(t) = I_0 T \cdot N(t)$$

$$S_I(\omega) = 4\pi I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \underbrace{\left[\tilde{N}(\omega - n\omega_T) \right]^2}_{\text{noise sidebands}} \right\}$$



noise sidebands



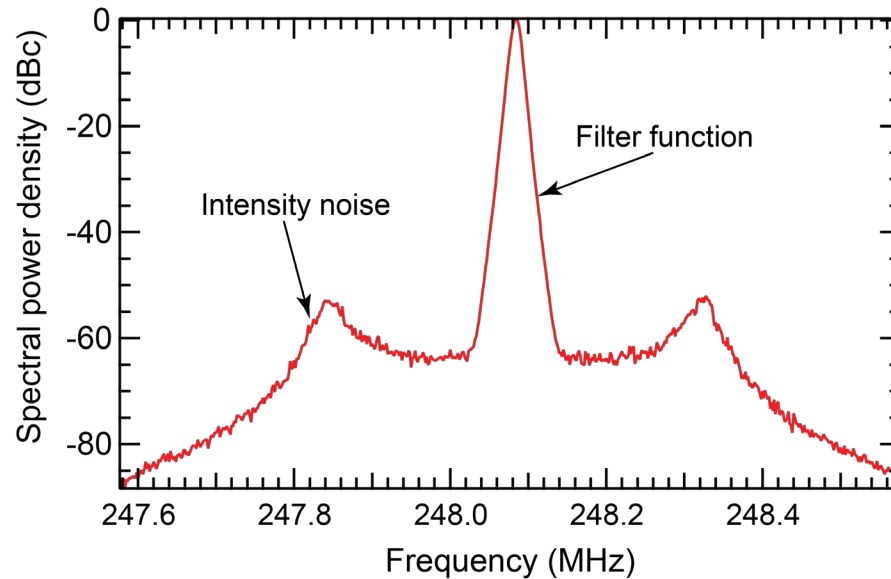


$$\text{dB} \equiv 10 \log \frac{P_2}{P_1}$$

$$\text{dBm} \equiv 10 \log \frac{P}{1 \text{ mW}}$$

mW	dBm
0.1 mW	-10 dBm
1 mW	0 dBm
10 mW	10 dBm
100 mW	20 dBm
1000 mW	30 dBm
x 2	+ 3 dBm
20 mW	13 dBm





dBc

“how many dB below the carrier”

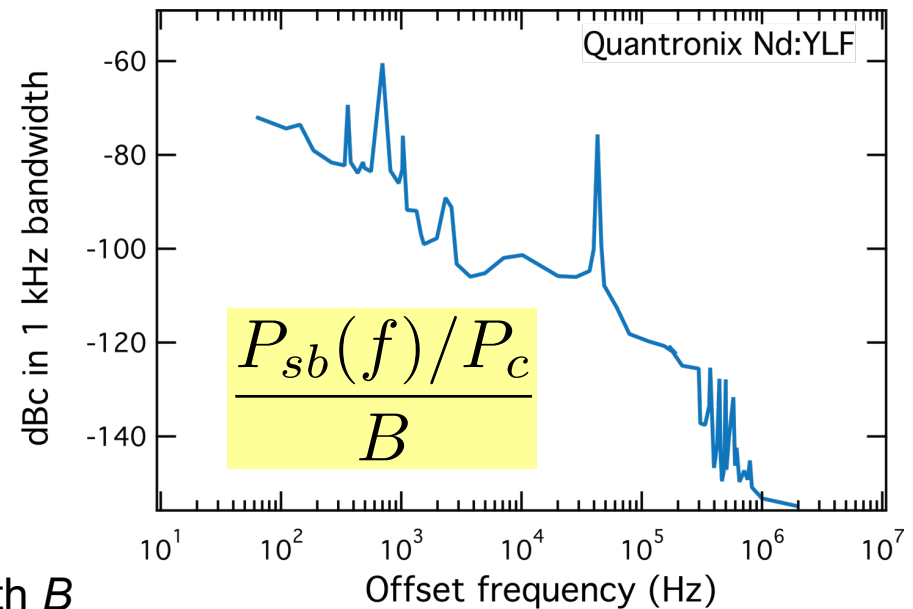
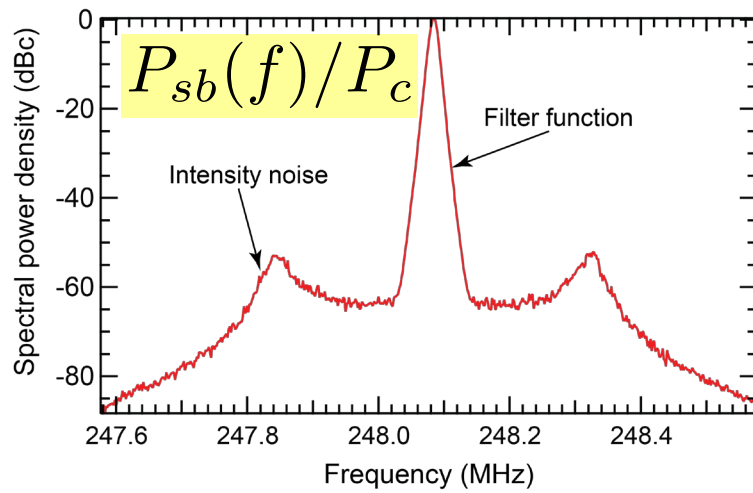
carrier is the peak of the harmonic signal

$$\text{dB} \equiv 10 \log \frac{P_2}{P_1}$$

$$\text{dBm} \equiv 10 \log \frac{P}{1 \text{ mW}}$$

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20 mW	13 dBm



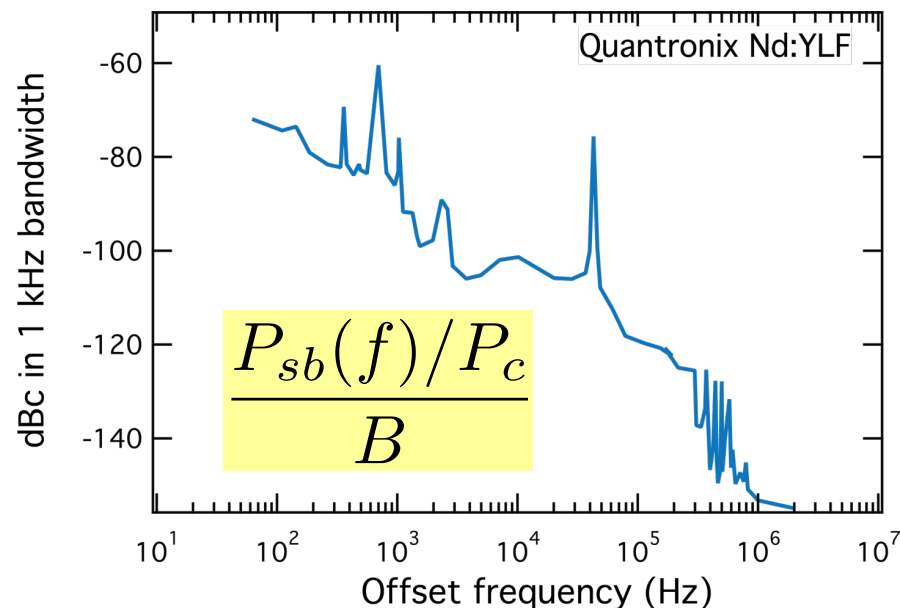


Measurement with a certain bandwidth B

- P_{sb} : power in the intensity sidebands (“two-sided” around harmonics)
- P_c : peak power of carrier
- f : offset frequency or noise frequency (i.e. deviation from laser harmonics)

$$\frac{P_{sb}}{P_c} = 2 \int_{nf_T + f_1}^{nf_T + f_2} \frac{P_{sb}(f)/P_c}{B} df$$

factor 2 because “two-sided” spectral density!



P_{sb} : power in the intensity sidebands

P_c : peak power of carrier

$$\frac{P_{sb}}{P_c} = 2 \int_{nf_T+f_1}^{nf_T+f_2} \frac{P_{sb}(f)/P_c}{B} df$$

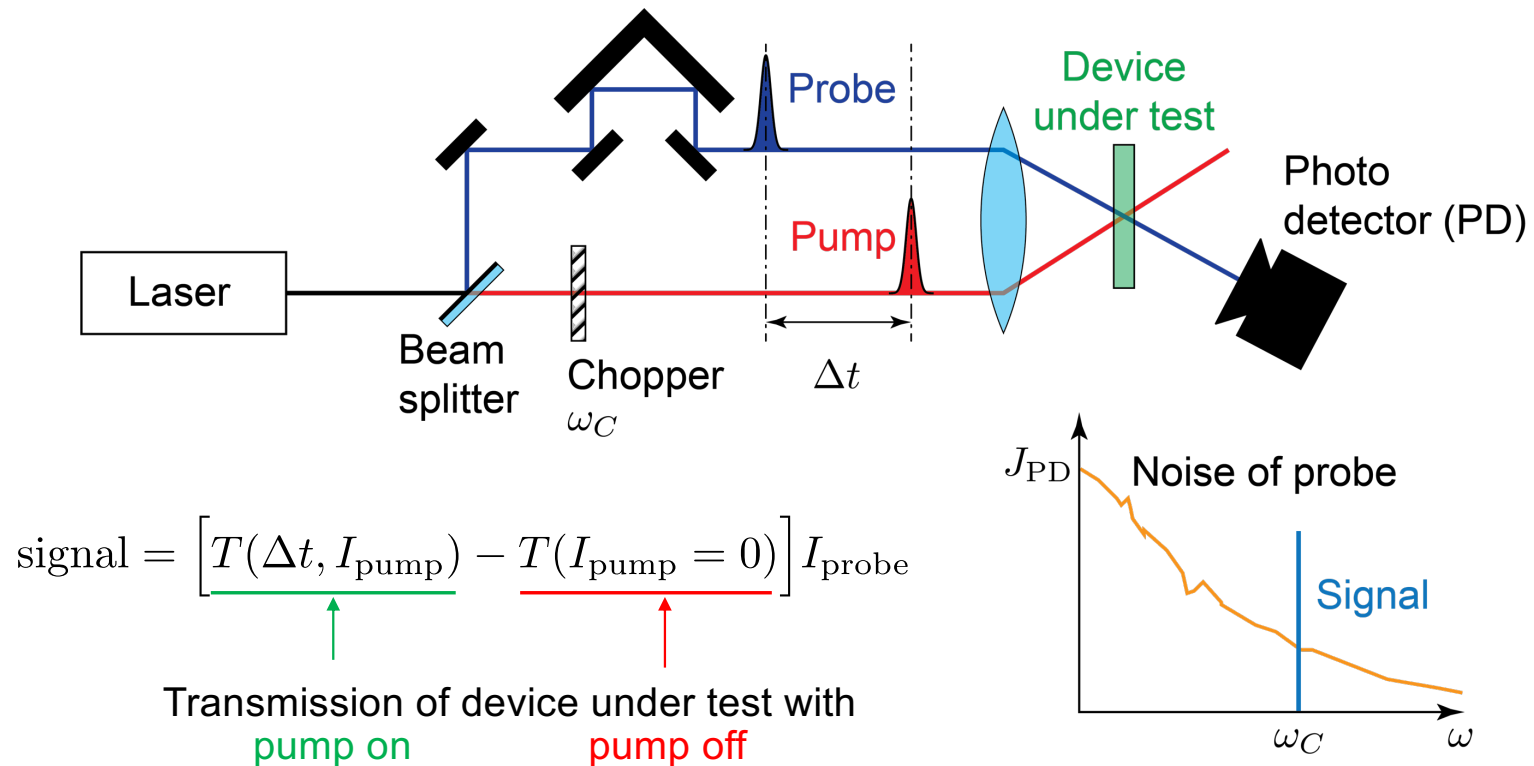
rms intensity noise or variance of intensity noise:

$$\begin{aligned} \sigma_N [\omega_1, \omega_2] &= \sqrt{\langle N^2(t) \rangle} \\ &= \sqrt{\frac{1}{\pi} \int_{\omega_1}^{\omega_2} S_N(\omega) d\omega} \end{aligned}$$

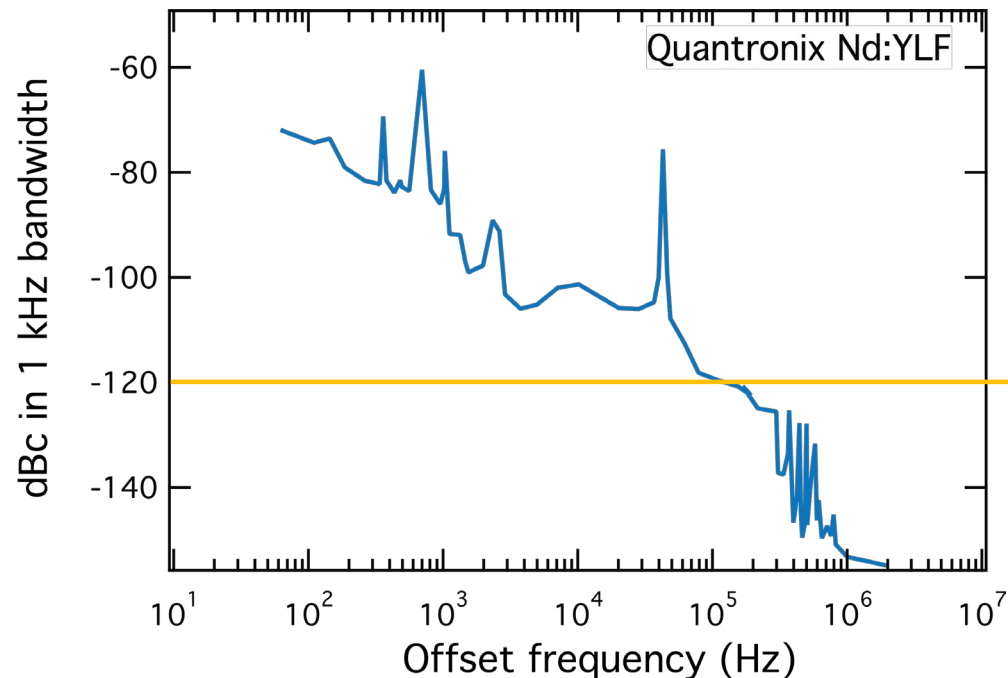


$$\sigma_N [f_1, f_2] = \sqrt{\frac{P_{sb}}{P_c}}$$

Differential transmission spectroscopy



Ultrafast measurements need some kind of nonlinearities in the measurement system (i.e. intensity dependent transmission)



Note: Microwave spectrum analyzer measures **power** of photocurrent!

$$20 \log \frac{S}{N} = 20 \log 10^{-6} = -120 \text{ dBc}$$

Need a chop frequency of > 100 kHz

$$I(t) = I_0 T \sum_{n=-\infty}^{+\infty} \delta(t - nT - \Delta T(t))$$



$$\tilde{I}_{\Delta T}(\omega) = F\{I_{\Delta T}(t)\} = -I_0 \sum_{n=-\infty}^{+\infty} in\omega_T \Delta\tilde{T}(\omega - n\omega_T)$$

$$S_{I_{\Delta T}}(\omega) = \tilde{I}_{\Delta T}(\omega) \cdot \tilde{I}_{\Delta T}^*(\omega) = I_0^2 (-i) \cdot (+i) \sum_{n=-\infty}^{+\infty} n^2 \omega_T^2 \left[\Delta\tilde{T}(\omega - n\omega_T) \right]^2$$



$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \frac{1}{4\pi^2} n^2 \omega_T^2 \left[\Delta\tilde{T}(\omega - n\omega_T) \right]^2 \right\}$$

$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \frac{1}{4\pi^2} n^2 \omega_T^2 \left[\Delta\tilde{T}(\omega - n\omega_T) \right]^2 \right\}$$

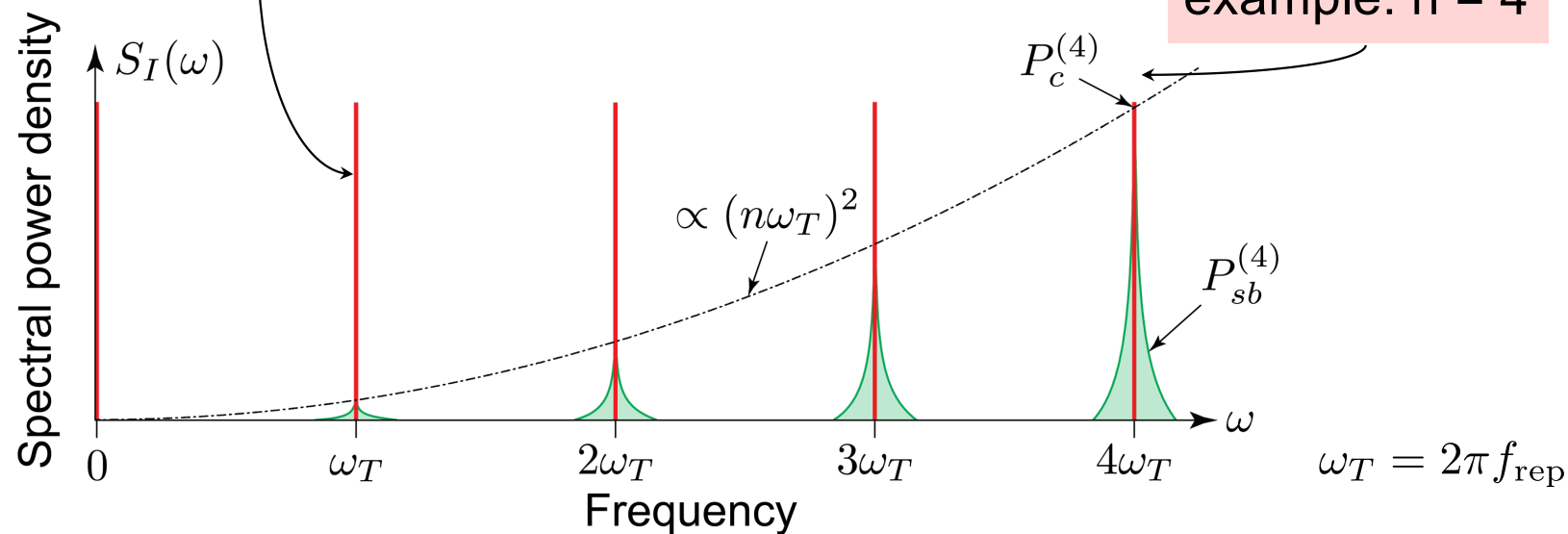
normalized power density
(two-sided around harmonics)

$$L_1(f) = \frac{P_{sb}^{(1)}}{P_c^{(1)}} = \frac{1}{n^2} L_n(f)$$

noise sidebands

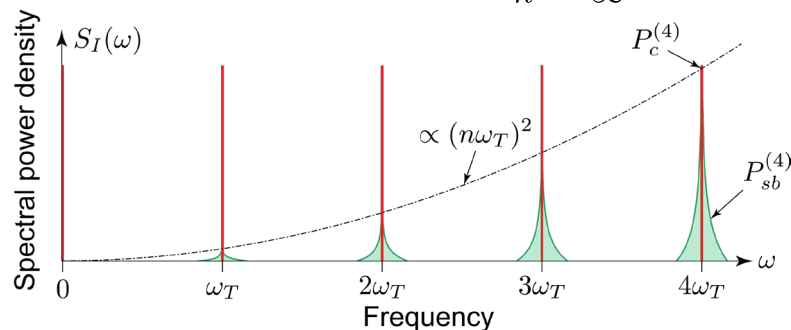
$$L_n(f) = \frac{P_{sb}^{(n)}}{P_c^{(n)}}$$

example: $n = 4$

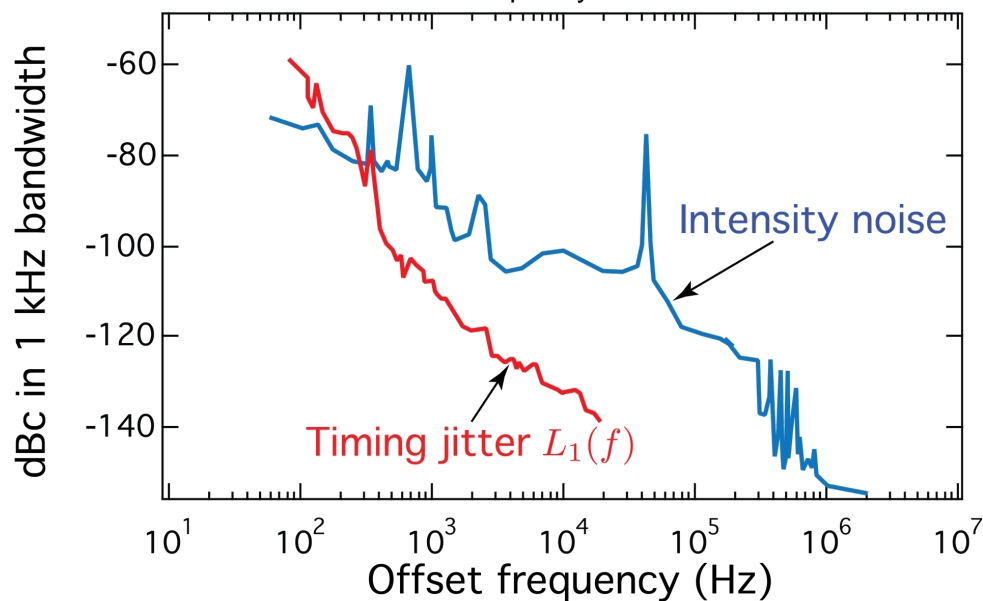


Timing jitter of mode-locked laser

$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \frac{1}{4\pi^2} n^2 \omega_T^2 \left[\Delta\tilde{T}(\omega - n\omega_T) \right]^2 \right\}$$



noise sidebands

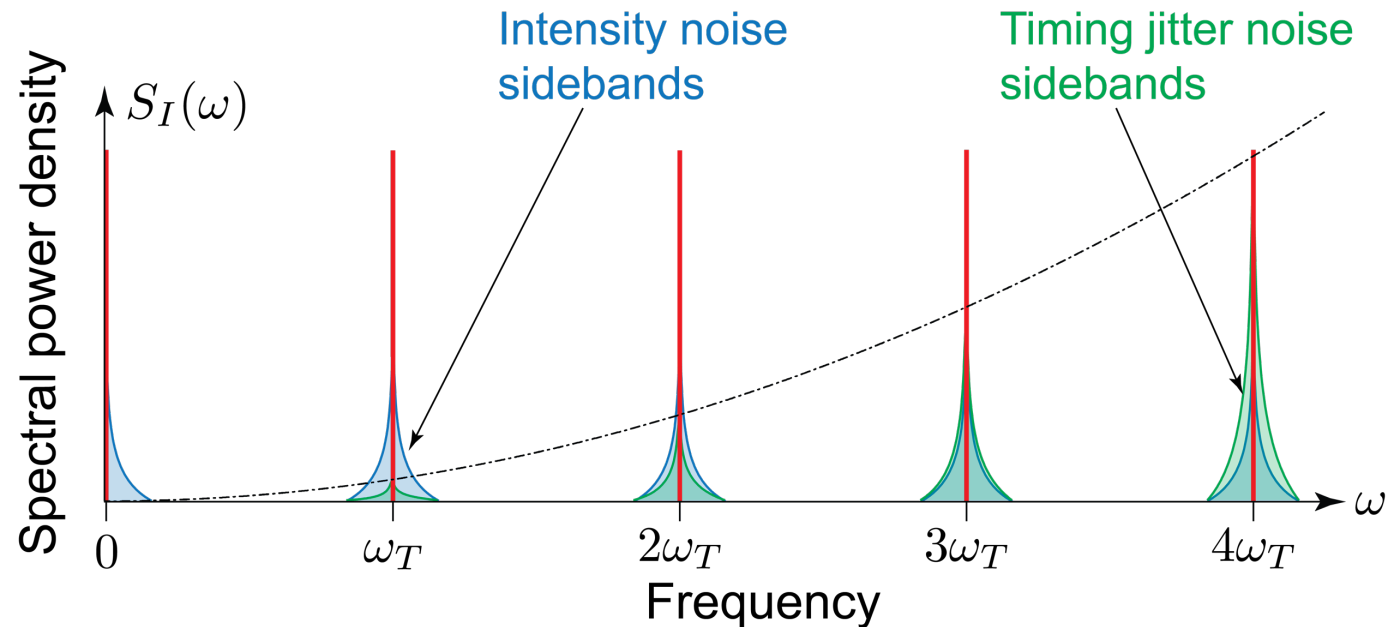


$$L_1(f) = \frac{P_{sb}^{(1)}}{P_c^{(1)}} = \frac{1}{n^2} L_n(f)$$

Measurement takes place at the n -th harmonic

Timing jitter sidebands stronger than intensity noise sidebands

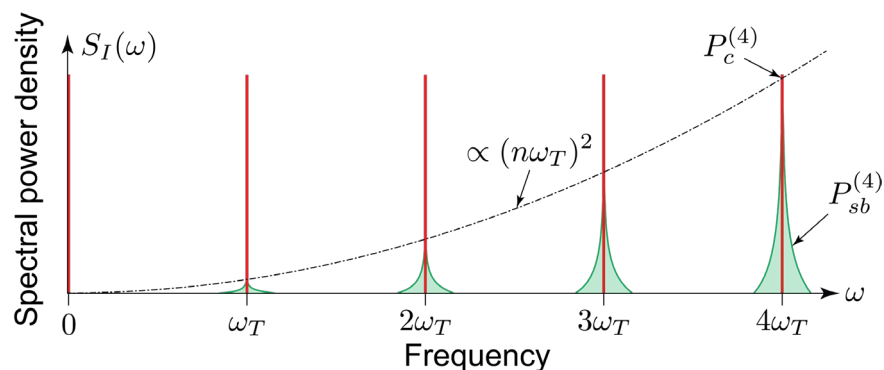
Double check: $\propto n^2$
small noise approximation



$$I(t) = I_0 T (1 + N(t)) \sum_{n=-\infty}^{+\infty} \delta(t - nT - \Delta T(t))$$

$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \left[\tilde{N}(\omega - n\omega_T) \right]^2 + \frac{1}{4\pi^2} n^2 \omega_T^2 \left[\Delta \tilde{T}(\omega - n\omega_T) \right]^2 \right\}$$

$$S_I(\omega) = 4\pi^2 I_0^2 \sum_{n=-\infty}^{+\infty} \left\{ \delta(\omega - n\omega_T) + \frac{1}{4\pi^2} n^2 \omega_T^2 \left[\Delta\tilde{T}(\omega - n\omega_T) \right]^2 \right\}$$



noise sidebands

$$L_1(f) = \frac{1}{n^2} L_n(f) = \frac{1}{n^2} \frac{P_{sb}^{(n)}}{P_c^{(n)}}$$

$$\sigma_{\Delta T} [f_1, f_2] = \sqrt{\langle \Delta T(t)^2 \rangle} = \frac{1}{2\pi f_T} \sqrt{2 \int_{f_1}^{f_2} L_1(f) df} = \frac{1}{2\pi n f_T} \sqrt{2 \int_{f_1}^{f_2} L_n(f) df}$$

factor of 2
because $L_1(f)$ is “two-sided”

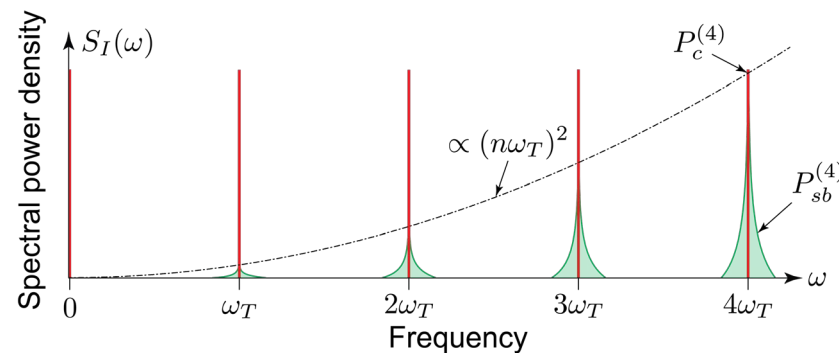
normalized power density
(two-sided around harmonics)

rms timing jitter:

$$\sigma_{\Delta T} [f_1, f_2] = \sqrt{\langle \Delta T(t)^2 \rangle} = \frac{1}{2\pi f_T} \sqrt{2 \int_{f_1}^{f_2} L_1(f) df} = \frac{1}{2\pi n f_T} \sqrt{2 \int_{f_1}^{f_2} L_n(f) df}$$

$$L_1(f) = \frac{P_{sb}^{(1)}(f)}{P_c^{(1)}}$$

$$L_n(f) = \frac{P_{sb}^{(n)}(f)}{P_c^{(n)}}$$



FWHM timing jitter:
(Gaussian noise statistics)

$$\Delta T(t) = \frac{1}{\sqrt{2\pi\sigma_{\Delta T}}} \exp\left(\frac{-t^2}{2\sigma_{\Delta T}^2}\right) = \exp\left(-4 \ln 2 \left(\frac{t}{\tau_{\Delta T}}\right)^2\right)$$

$$\tau_{\Delta T} = \sqrt{8 \ln 2} \sigma_{\Delta T} = 2.355 \cdot \sigma_{\Delta T}$$

Example:

$$\sigma_{\Delta T} [130 \text{ Hz}, 20 \text{ kHz}] \approx \sigma_{\Delta T} [130 \text{ Hz}, \infty] = 9 \text{ ps} \Rightarrow \approx 21 \text{ ps FWHM}$$